

TPS2112A TPS2113A

SBVS045C - MARCH 2004 - REVISED MAY 2012

# **AUTOSWITCHING POWER MUX**

Check for Samples: TPS2112A, TPS2113A

### FEATURES

- Two-Input, One-Output Power Multiplexer with Low r<sub>DS(on)</sub> Switches:
  - 84 mΩ Typ (TPS2113A)
  - 120 mΩ Typ (TPS2112A)
- Reverse and Cross-Conduction Blocking
- Wide Operating Voltage: 2.8 V to 5.5 V
- Low Standby Current: 0.5 µA Typ
- Low Operating Current: 55 µA Typ
- Adjustable Current Limit
- Controlled Output Voltage Transition Time:
  - Limits Inrush Current
  - Minimizes Output Voltage Hold-Up Capacitance
- CMOS- and TTL-Compatible Control Inputs
- Auto-Switching Operating Mode
- Thermal Shutdown
- Available in TSSOP-8 and 3-mm × 3-mm SON-8 Packages

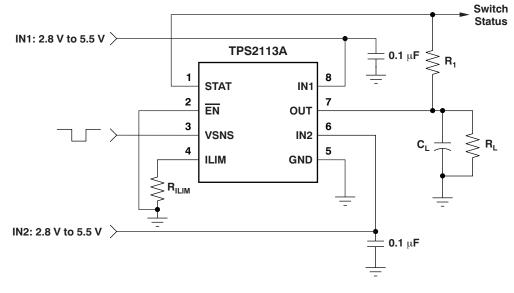
### **APPLICATIONS**

- PCs
- PDAs
- Digital Cameras
- Modems
- Cell Phones
- Digital Radios
- MP3 Players

### DESCRIPTION

The TPS211xA family of power multiplexers enables seamless transition between two power supplies (such as a battery and a wall adapter), each operating at 2.8 V to 5.5 V and delivering up to 2 A, depending on package. The TPS211xA family includes extensive protection circuitry, including user-programmable current limiting, thermal protection, inrush current control, seamless supply transition, cross-conduction blocking, and reverse-conduction blocking. These features greatly simplify designing power multiplexer applications.

### TYPICAL APPLICATION



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### **TPS2112A** TPS2113A



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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

		AVA	LABLE OPT	IONS			
FEATU	IRE	TPS2110A	TPS2111A	TPS2112A	TPS2113A	TPS2114A	TPS2115A
Current Limit Adjustment	t Range	0.31 A to 0.75 A	0.63 A to 1.25 A	0.31 A to 0.75 A	0.63 A to 2 A	0.31 A to 0.75 A	0.63 A to 2 A
Switching Modeo	Manual	Yes	Yes	No	No	Yes	Yes
Switching Modes	Automatic	Yes	Yes	Yes	Yes	Yes	Yes
Switch Status Output		No	No	Yes	Yes	Yes	Yes

#### **DEVICE INFORMATION<sup>(1)</sup>**

T <sub>A</sub>	PACKAGE	I <sub>OUT</sub> (A)	ORDERING NUMBER	PACKAGE MARKING		
−40°C to +85°C		0.75	TPS2112APW	2112A		
	TSSOP-8 (PW)	1.25	TPS2113APW	2112A 2113A		
	SON-8 (DRB)	2	TPS2113ADRB	PTOI		

For the most current package and ordering information see the Package Option Addendum at the end of this document, or see the TI (1)web site at www.ti.com.

### **ABSOLUTE MAXIMUM RATINGS<sup>(1)</sup>**

Over recommended junction temperature range, unless otherwise noted.

			TPS2112A, TPS2113A	UNIT	
Input vo	oltage range at pins IN1, IN	N2, EN, VSNS, ILIM <sup>(2)</sup>	-0.3 to 6	V	
Output voltage range, V <sub>O(OUT)</sub> , V <sub>O(STAT)</sub> <sup>(2)</sup>			-0.3 to 6	V	
Output	sink current, I <sub>O(STAT)</sub>		5		
		TPS2112APW	0.9	А	
Continu	ious output current, I <sub>O</sub>	TPS2113APW	1.5	А	
		TPS2113ADRB, T <sub>J</sub> ≤ 105°C	2.5	А	
Continu	ious total power dissipatior	n	See Dissipation Ratin	igs table	
Junctio	n temperature		Internally Limite	ed	
	Human body model (HE	3M)	2	kV	
ESD	Charged device model (	Charged device model (CDM)		V	

(1) Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

All voltages are with respect to GND. (2)

#### **DISSIPATION RATINGS**

PACKAGE	DERATING FACTOR ABOVE T <sub>A</sub> = 25°C	T <sub>A</sub> ≤ 25°C POWER RATING	T <sub>A</sub> = 70°C POWER RATING	T <sub>A</sub> = 85°C POWER RATING
TSSOP-8 (PW)	3.9 mW/°C	387 mW	213 mW	155 mW
SON-8 (DRB) <sup>(1)</sup>	25.0 mW/°C	2.50 mW	1.38 mW	1.0 W

(1) See TI application note SLMA002 for mounting recommendations.

### **RECOMMENDED OPERATING CONDITIONS**

		TPS2112A, TPS2113A			
		MIN	NOM MAX	UNIT	
Input veltage et IN4 V	V <sub>I(IN2)</sub> ≥ 2.8 V	1.5	5.5	V	
Input voltage at IN1, $V_{I(IN1)}$	V <sub>I(IN2)</sub> < 2.8 V	2.8	5.5	V	
Input voltage et IN2 V	V <sub>I(IN1)</sub> ≥ 2.8 V	1.5	5.5	V	
Input voltage at IN2, $V_{I(IN2)}$	V <sub>I(IN1)</sub> < 2.8 V	2.8	5.5	v	
Input voltage: V <sub>I(EN)</sub> , V <sub>I(VSNS)</sub>		0	5.5	V	
	TPS2112APW	0.31	0.75	٨	
Nominal current limit adjustment range, $I_{O(OUT)}^{(1)}$	TPS2113APW	0.63	1.25	A	
	TPS2113ADRB, T <sub>J</sub> ≤ 105°C	0.63	2	А	
Operating virtual junction temperature, T	J	-40	125	°C	

(1) Minimum recommended current limit is based on accuracy considerations.

### ELECTRICAL CHARACTERISTICS: Power Switch

Over recommended operating junction temperature,  $R_{ILIM}$  = 400  $\Omega$ , unless otherwise noted.

				TF	S2112A		TP	S2113A			
PARAME	PARAMETER TEST CONDITIO		CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	UNIT	
2				$V_{I(IN1)} = V_{I(IN2)} = 5.0 V$		120	140		84	110	
		$T_J = 25^{\circ}C,$ $I_I = 500 \text{ mA}$	$V_{I(IN1)} = V_{I(IN2)} = 3.3 V$		120	140		84	110	mΩ	
Drain-source on-state	- (1)		V <sub>I(IN1)</sub> = V <sub>I(IN2)</sub> = 2.8 V		120	140		84	110		
resistance (INx-OUT)	r <sub>DS(on)</sub> <sup>(1)</sup>		$V_{I(IN1)} = V_{I(IN2)} = 5.0 V$			220			150		
(INX-001)		T <sub>J</sub> = 125°C, I <sub>I</sub> = 500 mA	$V_{I(IN1)} = V_{I(IN2)} = 3.3 V$			220			150	mΩ	
		12 - 5		$V_{I(IN1)} = V_{I(IN2)} = 2.8 V$			220			150	

(1) The TPS211xA can switch a voltage as low as 1.5 V as long as there is a minimum of 2.8 V at one of the input power pins. In this specific case, the lower supply voltage has no effect on the IN1 and IN2 switch on-resistances.

### **ELECTRICAL CHARACTERISTICS**

Over recommended operating junction temperature,  $I_{O(OUT)} = 0$  A, and  $R_{ILIM} = 400 \Omega$ , unless otherwise noted.

			TPS2112A, TPS2113A				
PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
LOGIC INPUTS (EN)							
High-level input voltage	VIH		2			V	
Low-level input voltage	V <sub>IL</sub>				0.7	V	
land an entry		EN = High, sink current			1		
Input current		EN = Low, source current	0.5	1.4	5	μA	
SUPPLY AND LEAKAGE	CURRENTS						
		$\label{eq:VIVSNS} \begin{array}{l} V_{I(VSNS)} = 1.5 \ V, \ \overline{EN} = Low \ (IN1 \ active), \\ V_{I(IN1)} = 5.5 \ V, \ V_{I(IN2)} = 3.3 \ V \end{array}$		55	90		
Supply autoat from INI4 (a	noroting)	$\label{eq:VIVSNS} \begin{array}{l} V_{I(VSNS)} = 1.5 \ V, \ \overline{\text{EN}} = \text{Low (IN1 active)}, \\ V_{I(IN1)} = 3.3 \ V, \ V_{I(IN2)} = 5.5 \ V, \end{array}$	1		12	μA	
Supply current from IN1 (o	perating)				75	μΛ	
		$\label{eq:VIVSNS} \begin{array}{l} V_{I(VSNS)} = 0 \ V, \ \overline{EN} = Low \ (IN2 \ active), \\ V_{I(IN1)} = 3.3 \ V, \ V_{I(IN2)} = 5.5 \ V \end{array}$			1		
		$\label{eq:VIVSNS} \begin{array}{l} V_{I(VSNS)} = 1.5 \ V, \ \overline{EN} = Low \ (IN1 \ active), \\ V_{I(IN1)} = 5.5 \ V, \ V_{I(IN2)} = 3.3 \ V \end{array}$			1		
Supply autoat from IND (a	noroting)	$V_{I(VSNS)} = 1.5 \text{ V}, \overline{EN} = \text{Low (IN1 active)},$ $V_{I(IN1)} = 3.3 \text{ V}, V_{I(IN2)} = 5.5 \text{ V}$			75		
Supply current from IN2 (o	peraung)	$\label{eq:VIVSNS} \begin{array}{l} V_{I(VSNS)} = 0 \ V, \ \overline{\text{EN}} = \text{Low} \ (\text{IN2 active}), \\ V_{I(\text{IN1})} = 5.5 \ V, \ V_{I(\text{IN2})} = 3.3 \ V \end{array}$		1	12	μA	
		$\label{eq:VIVSNS} \begin{array}{l} V_{I(VSNS)} = 0 \ V, \ \overline{EN} = Low \ (IN2 \ active), \\ V_{I(IN1)} = 3.3 \ V, \ V_{I(IN2)} = 5.5 \ V \end{array}$		55	90		

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### **ELECTRICAL CHARACTERISTICS (continued)**

Over recommended operating junction temperature,  $I_{O(OUT)} = 0$  A, and  $R_{ILIM} = 400 \Omega$ , unless otherwise noted.

			TPS2112A, TPS2113		3A	
PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
SUPPLY AND LEAKAGE	CURRENTS, Cor	ntinued				
0		$\overline{\text{EN}} = \text{High (inactive), V}_{I(\text{IN1})} = 5.5 \text{ V},$ V <sub>1(IN2)</sub> = 3.3 V			2	
Quiescent current from IN1	(standby)	$\overline{\text{EN}}$ = High (inactive), V <sub>I(IN1)</sub> = 3.3 V, V <sub>I(IN2)</sub> = 5.5 V			1	μA
Ouissaant surrant from IN2	(ato adday)	$\overline{\text{EN}} = \text{High (inactive), V}_{\text{I(IN1)}} = 5.5 \text{ V},$ V <sub>I(IN2)</sub> = 3.3 V			1	
Quiescent current from IN2	(standby)	$\overline{\text{EN}} = \text{High (inactive)}, \text{ V}_{\text{I(IN1)}} = 3.3 \text{ V}, \\ \text{V}_{\text{I(IN2)}} = 5.5 \text{ V}$		0.5	2	μA
Forward leakage current fro (measured from OUT to GN		$\label{eq:constraint} \begin{array}{l} \overline{\text{EN}} = \text{High (inactive)}, \ \text{V}_{\text{I(IN1)}} = 5.5 \ \text{V}, \ \text{IN2 open}, \\ \text{V}_{\text{O(OUT)}} = 0 \ \text{V} \ \text{(shorted)}, \ \text{T}_{\text{J}} = 25^{\circ}\text{C} \end{array}$		0.1	5	μA
Forward leakage current fro (measured from OUT to GN		$\label{eq:Volume} \begin{array}{l} \overline{\text{EN}} = \text{High (inactive)}, \ \text{V}_{\text{I(IN2)}} = 5.5 \ \text{V}, \ \text{IN1 open}, \\ \text{V}_{\text{O(OUT)}} = 0 \ \text{V} \ \text{(shorted)}, \ \text{T}_{\text{J}} = 25^{\circ}\text{C} \end{array}$		0.1	5	μA
Reverse leakage current to from INx to GND)	INx (measured	$\label{eq:constraint} \begin{array}{l} \overline{\text{EN}} = \text{High (inactive)}, \ \text{V}_{\text{I(INx)}} = 0 \ \text{V}, \\ \text{V}_{\text{O(OUT)}} = 5.5 \ \text{V}, \ \text{T}_{\text{J}} = 25^{\circ}\text{C} \end{array}$		0.3	5	μA
STAT OUTPUT						
Leakage current		V <sub>O(STAT)</sub> = 5.5 V		0.01	1	μA
Saturation voltage		$I_{I(STAT)} = 2 \text{ mA}$ , IN1 switch is on		0.13	0.4	V
Deglitch time (falling edge only)				150		μs
CURRENT LIMIT CIRCUIT						
	TPS2112A	$R_{ILIM} = 400 \ \Omega$	0.51	0.63	0.80	A 0 A
Current limit accuracy	11 32112A	R <sub>ILIM</sub> = 700 Ω	0.30	0.36	0.50	
	TPS2113A	$R_{ILIM} = 400 \ \Omega$	0.95	1.25	1.56	
	IF 52115A	R <sub>ILIM</sub> = 700 Ω	0.47	0.71	0.99	
Current limit settling time	t <sub>d</sub>	Time for short-circuit output current to settle within 10% of its steady state value.		1		ms
Input current at ILIM		$V_{I(ILIM)} = 0 V$	-15		0	μA
VSNS COMPARATOR						
VENE throughold voltage		V <sub>I(VSNS)</sub> ↑	0.78	0.80	0.82	V
VSNS threshold voltage		$V_{I(VSNS)}\downarrow$	0.735	0.755	0.775	v
VSNS comparator hysteres	is		30		60	mV
Deglitch of VSNS comparate	tor (both $\uparrow\downarrow$ )		90	150	220	μs
Input current		$0 \text{ V} \leq \text{V}_{\text{I}(\text{VSNS})} \leq 5.5 \text{ V}$	-1		1	μA
UVLO						
		Falling edge	1.15	1.25		V
IN1 and IN2 UVLO		Rising edge		1.30	1.35	v
IN1 and IN2 UVLO hysteres	sis		30	57	65	mV
Internal V <sub>DD</sub> UVLO (the hig	her of IN1 and	Falling edge	2.4	2.53		V
IN2)		Rising edge		2.58	2.8	v
Internal V <sub>DD</sub> UVLO hysteres	sis		30	50	75	mV
UVLO deglitch for IN1, IN2		Falling edge		110		μs



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### **ELECTRICAL CHARACTERISTICS (continued)**

Over recommended operating junction temperature,  $I_{O(OUT)} = 0$  A, and  $R_{ILIM} = 400 \Omega$ , unless otherwise noted.

			TPS211	2A, TPS211:	BA	
PARAMETER		TEST CONDITIONS	MIN	TYP	МАХ	UNIT
REVERSE CONDUCTION	BLOCKING					
Minimum output-to-input voltage difference to block switching	$\Delta V_{O(I\_block)}$	$\label{eq:response} \begin{array}{ c c c c } \hline \overline{EN} = high, \ V_{I(IN1)} = 3.3 \ V \ and \ V_{I(IN2)} = V_{I(VSNS)} \\ = 0 \ V. \ Connect \ OUT \ to \ a \ 5-V \ supply \ through \\ a \ series \ 1-k\Omega \ resistor. \ Let \ \overline{EN} = low. \ Slowly \\ decrease \ the \ supply \ voltage \ until \ OUT \\ connects \ to \ IN1. \end{array}$	80	100	120	mV
THERMAL SHUTDOWN						
Thermal shutdown threshold	Ł	TPS211xA is in current limit.	135			°C
Recovery from thermal shut	down	TPS211xA is in current limit.	125			°C
Hysteresis				10		°C
IN2-IN1 COMPARATORS						
Hysteresis of IN2-IN1 comp	arator		0.1		0.2	V
Deglitch of IN2-IN1 compar	ator (both $\uparrow \downarrow$ )		10	20	50	μs

### SWITCHING CHARACTERISTICS

Over recommended operating junction temperature,  $V_{I(IN1)} = V_{I(IN2)} = 5.5$  V, and  $R_{ILIM} = 400 \Omega$ , unless otherwise noted.

				TPS2112A		TPS2113A				
Р	ARAMETER	TEST CON	DITIONS	MIN	ТҮР	MAX	MIN TYP		MAX	UNIT
t <sub>R</sub>	Output rise time from an enable		$\label{eq:constraint} \begin{array}{l} T_J = 25^\circ\text{C},\\ C_L = 1\ \mu\text{F},\\ I_L = 500\ \text{mA};\ \text{see}\\ \hline \text{Figure 1(a)}. \end{array}$	0.5	1.0	1.5	1	1.8	3	ms
t <sub>F</sub>	Output fall time from a disable	$ \begin{array}{l} V_{l(IN1)} = V_{l(IN2)} = 5 \ V, \\ V_{l(SNS)} = 1.5 \ V \end{array} $	$ \begin{array}{l} T_J = 25^\circ C, \\ C_L = 1 \ \mu F, \\ I_L = 500 \ m A; \ see \\ Figure \ 1(a). \end{array} $	0.35	0.5	0.7	0.5	1	2	ms
t <sub>T</sub>	Transition time		$\begin{array}{l} T_J = 125^\circ C,\\ C_L = 10 \ \mu F,\\ I_L = 500 \ mA; \ measure\\ transition time as 10\%\\ to 90\% \ rise time or\\ from 3.4 \ V to 4.8 \ V on\\ V_{O(OUT)}. \ See\\ Figure 1 (b). \end{array}$		40	60		40	60	μs
t <sub>PLH1</sub>	Turn-on propagation delay from an enable	$ \begin{array}{l} V_{I(IN1)} = VI_{(IN2)} = 5 \ V \\ Measured from enable to \\ 10\% \ of \ V_{O(OUT)}, \ V_{I(SNS)} = \\ 1.5 \ V \end{array} $	$ \begin{array}{l} T_{J} = 25^{\circ}C, \\ C_{L} = 10 \ \mu\text{F}, \\ I_{L} = 500 \ \text{mA}; \ \text{see} \\ \hline \text{Figure 1(a)}. \end{array} $		0.5			1		ms
t <sub>PHL1</sub>	Turn-off propagation delay from a disable	$ \begin{array}{l} V_{l(IN1)} = VI_{(IN2)} = 5 \ V \\ \mbox{Measured from disable to} \\ 90\% \ of \ V_{O(OUT)}, \ V_{l(SNS)} = \\ 1.5 \ V \end{array} $	$T_J = 25^{\circ}C,$ $C_L = 10 \ \mu F,$ $I_L = 500 \ mA;$ see Figure 1(a).		3			5		ms
t <sub>PLH2</sub>	Switch-over rising propagation delay	$\begin{array}{l} \mbox{Logic 1 to Logic 0} \\ \mbox{transition on VSNS,} \\ V_{I(IN1)} = 1.5 V, \\ V_{I(IN2)} = 5 V, \\ V_{I(\overline{EN})} = 0 V, \\ \mbox{Measured from VSNS to} \\ \mbox{10\% of } V_{O(OUT)} \end{array}$	$\begin{array}{l} T_{J} = 25^{\circ}C, \\ C_{L} = 10 \ \mu\text{F}, \\ I_{L} = 500 \ \text{mA}; \ \text{see} \\ Figure \ 1(c). \end{array}$		40	100		40	100	μs
t <sub>PHL2</sub>	Switch-over falling propagation delay	$\begin{array}{l} \mbox{Logic 0 to Logic 1} \\ \mbox{transition on VSNS,} \\ V_{I(IN1)} = 1.5 V, \\ V_{I(IN2)} = 5 V, \\ V_{I(IN2)} = 0 V, \\ \mbox{Measured from VSNS to} \\ \mbox{90\% of } V_{O(OUT)} \end{array}$	$\begin{array}{l} T_{\rm J} = 25^{\circ}C, \\ C_{\rm L} = 10 \ \mu\text{F}, \\ I_{\rm L} = 500 \ \text{mA}; \ \text{see} \\ Figure \ 1 (c). \end{array}$	2	3	10	2	5	10	ms

TPS2112A TPS2113A

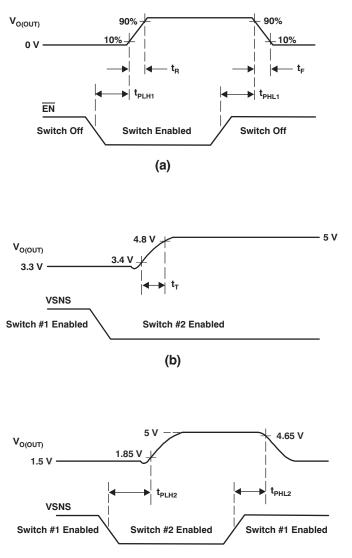
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PARAMETER MEASUREMENT INFORMATION

### TIMING WAVEFORMS



(C)

Figure 1. Propagation Delays and Transition Timing Waveforms



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#### **DEVICE INFORMATION**

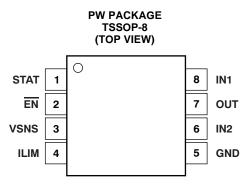
#### TRUTH TABLE

EN	$V_{I(VSNS)} > 0.8 V^{(1)}$	$V_{I(IN2)} > V_{I(IN1)}$	STAT	OUT <sup>(2)</sup>
0	Yes	Х	0	IN1
0	No	No	0	IN1
0	No	Yes	Hi-Z	IN2
1	Х	Х	0	Hi-Z

(1) X = Don't care.

(1) X = Donrotate.
 (2) The undervoltage lockout circuit causes the output (OUT) to go Hi-Z if the selected power supply does not exceed the IN1/IN2 UVLO, or if neither of the supplies exceeds the internal V<sub>DD</sub> UVLO.

### **PIN CONFIGURATIONS**



DRB PACKAGE SON-8 (TOP VIEW)							
STAT	1		8	IN1			
EN	2		7	OUT			
VSNS	3	GND	6	IN2			
ILIM	4		5	GND			

#### Table 1. TERMINAL FUNCTIONS

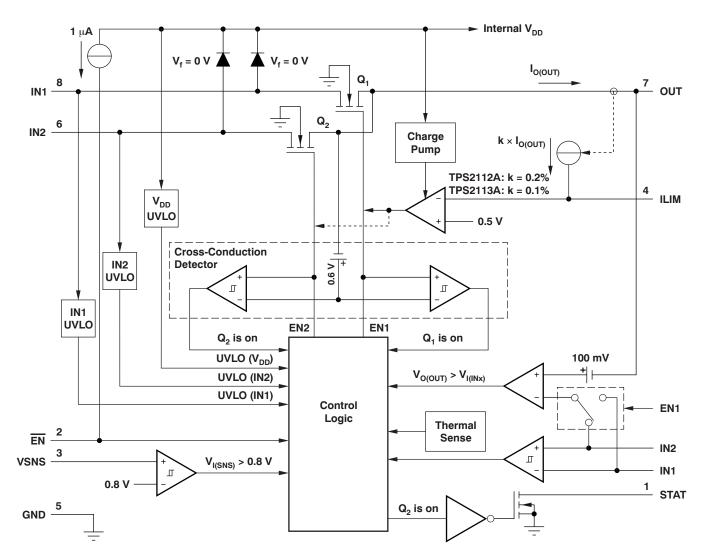
TERMINAL			
NAME	NO.	I/O	DESCRIPTION
ĒN	2	I	TTL- and CMOS-compatible input with a 1- $\mu$ A pull-up. The Truth Table illustrates the functionality of EN.
GND	5	Power	Ground
IN1	8	I	Primary power switch input. The IN1 switch can be enabled only if the IN1 supply is above the UVLO threshold and at least one supply exceeds the internal $V_{DD}$ UVLO.
IN2	6	I	Secondary power switch input. The IN2 switch can be enabled only if the IN2 supply is above the UVLO threshold and at least one supply exceeds the internal V <sub>DD</sub> UVLO.
ILIM	4	I	A resistor (R <sub>ILIM</sub> ) from ILIM to GND sets the current limit (I <sub>L</sub> ) to $250/R_{ILIM}$ and $500/R_{ILIM}$ for the TPS2112A and TPS2113A, respectively.
OUT	7	0	Power switch output
STAT	1	0	STAT is an open-drain output that is Hi-Z if the IN2 switch is ON. STAT pulls low if the IN1 switch is ON or if OUT is Hi-Z (that is, EN is equal to logic '0')
VSNS	3	I	An internal power FET connects OUT to IN1 if the VSNS voltage is greater than 0.8 V. Otherwise, the FET connects OUT to the higher of IN1 and IN2. The Truth Table illustrates the functionality of VSNS.
Pad	_	Power	<b>DRB package only.</b> Connect to GND. Must be connected to large copper area in order to meet stated package dissipation ratings.

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### FUNCTIONAL BLOCK DIAGRAM





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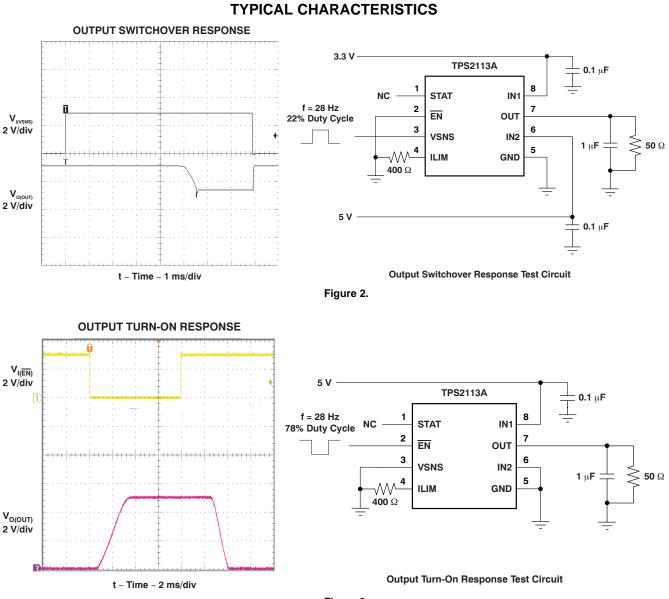


Figure 3.

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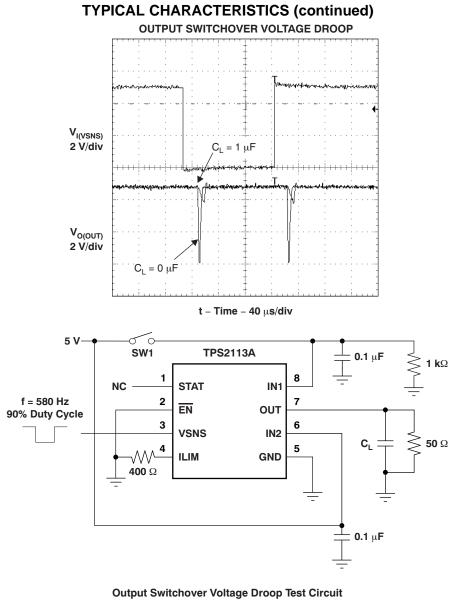
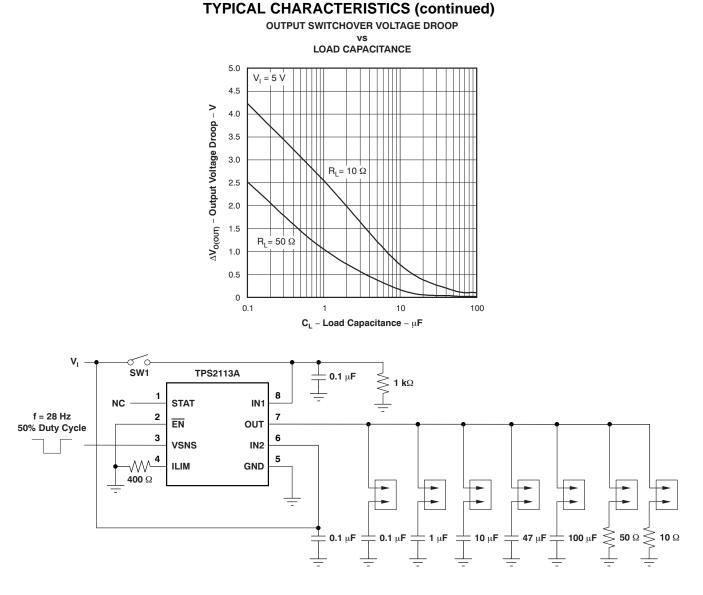


Figure 4.

Note: To initialize the TPS2113A for this test, set input VSNS equal to 0 V, turn on the 5-V supply, and then turn on switch SW1.



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**Output Switchover Voltage Droop Test Circuit** 

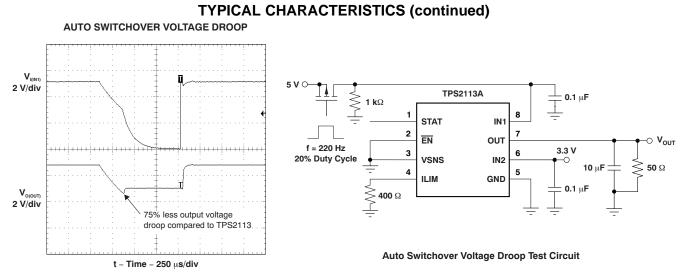
Figure 5.

Note: To initialize the TPS2113A for this test, set input VSNS equal to 0 V, turn on the  $V_1$  supply, and then turn on switch SW1.

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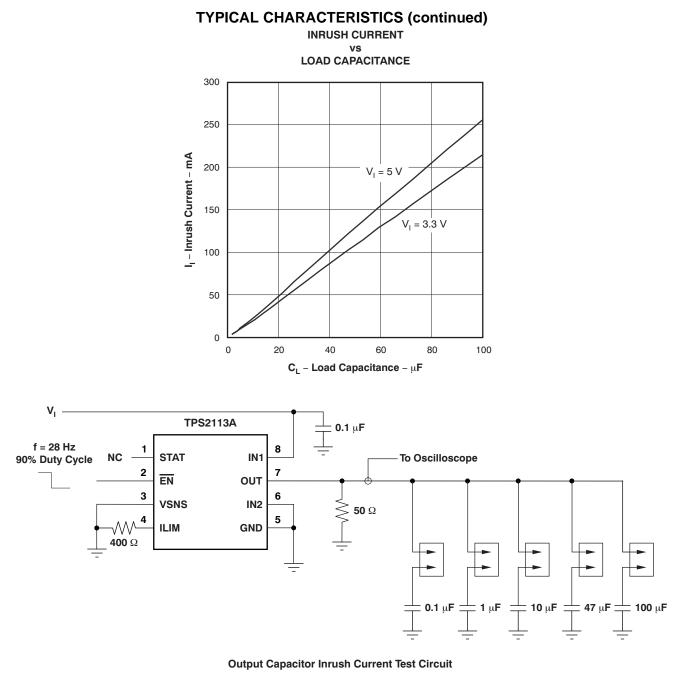
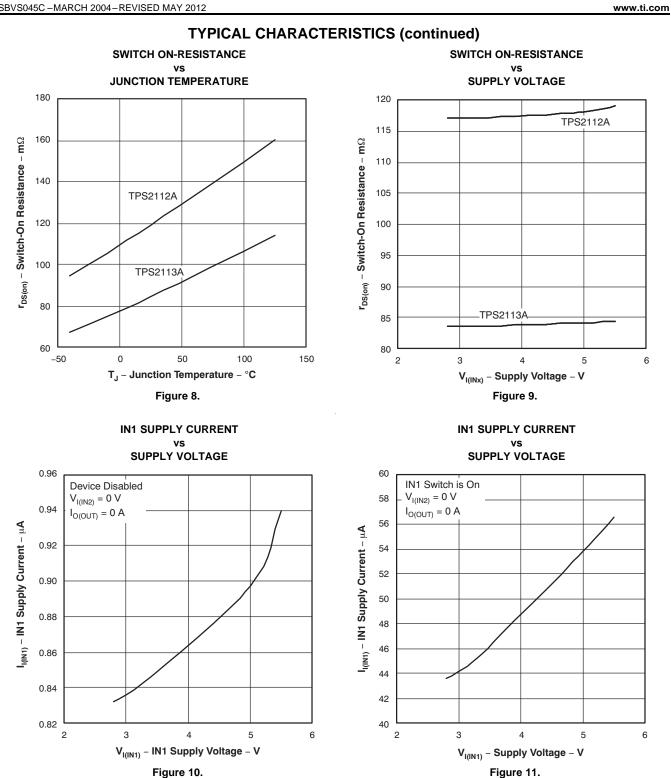


Figure 7.

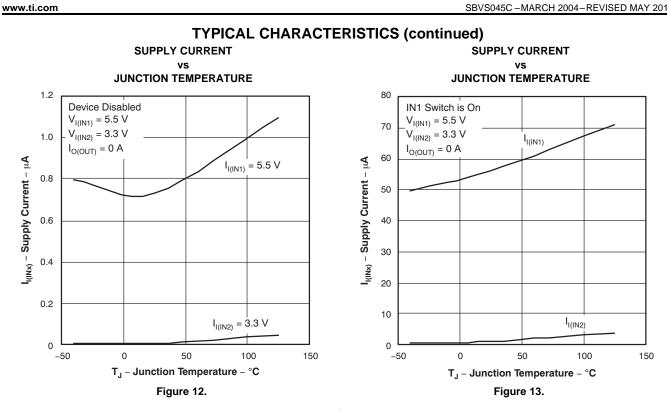
Texas NSTRUMENTS

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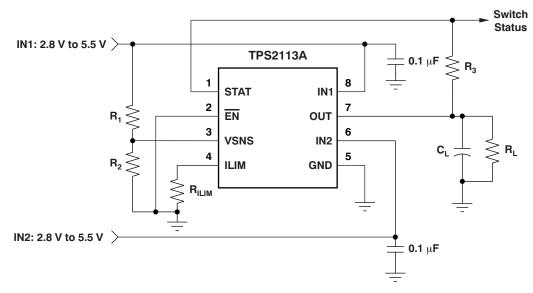




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### **APPLICATION INFORMATION**

Some applications have two energy sources, one of which should be used in preference to another. Figure 14 shows a circuit that will connect IN1 to OUT until the voltage at IN1 falls below a user-specified value. Once the voltage on IN1 falls below this value, the TPS2112A/3A will select the higher of the two supplies. This usually means that the TPS2112A/3A will swap to IN2.





In Figure 15, the multiplexer selects between two power supplies based upon the VSNS logic signal. OUT connects to IN1 if VSNS is logic '1'; otherwise, OUT connects to IN2 if  $V_{IN2}$  is greater than  $V_{IN1}$ . The logic thresholds for the VSNS terminal are compatible with both TTL and CMOS logic.

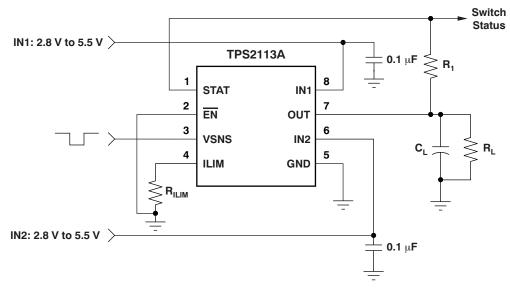


Figure 15. Manually Switching Power Sources



#### DETAILED DESCRIPTION

The TPS2112A/3A only supports the auto-switching mode. In this mode, OUT connects to IN1 if  $V_{I(VSNS)}$  is greater than 0.8 V, otherwise OUT connects to the higher of IN1 and IN2.

The VSNS terminal includes hysteresis equal to 3.75% to 7.5% of the threshold selected for transition from the primary supply to the higher of the two supplies. This hysteresis helps avoid repeated switching from one supply to the other due to resistive drops.

#### N-CHANNEL MOSFETs

Two internal high-side power MOSFETs implement a single-pole double-throw (SPDT) switch. Digital logic selects the IN1 switch, IN2 switch, or no switch (Hi-Z state). The MOSFETs have no parallel diodes so output-to-input current cannot flow when the FET is off. An integrated comparator prevents turn-on of a FET switch if the output voltage is greater than the input voltage.

#### **CROSS-CONDUCTION BLOCKING**

The switching circuitry ensures that both power switches will never conduct at the same time. A comparator monitors the gate-to-source voltage of each power FET and allows a FET to turn on only if the gate-to-source voltage of the other FET is below the turn-on threshold voltage.

#### **REVERSE-CONDUCTION BLOCKING**

When the TPS211xA switches from a higher-voltage supply to a lower-voltage supply, current can potentially flow back from the load capacitor into the lower-voltage supply. To minimize such reverse conduction, the TPS211xA will not connect a supply to the output until the output voltage has fallen to within 100 mV of the supply voltage. Once a supply has been connected to the output, it will remain connected regardless of output voltage.

#### CHARGE PUMP

The higher of supplies IN1 and IN2 powers the internal charge pump. The charge pump provides power to the current limit amplifier and allows the output FET gate voltage to be higher than the IN1 and IN2 supply voltages. A gate voltage that is higher than the source voltage is necessary to turn on the N-channel FET.

#### CURRENT LIMITING

A resistor  $R_{ILIM}$  from ILIM to GND sets the current limit to 250/ $R_{ILIM}$  and 500/ $R_{ILIM}$  for the TPS2112A and TPS2113A, respectively. Setting resistor  $R_{ILIM}$  equal to zero is not recommended as that disables current limiting.

#### **OUTPUT VOLTAGE SLEW-RATE CONTROL**

The TPS2112A/3A slews the output voltage at a slow rate when OUT switches to IN1 or IN2 from the Hi-Z state (see the Truth Table). A slow slew rate limits the inrush current into the load capacitor. High inrush currents can glitch the voltage bus and cause a system to hang up or reset. It can also cause reliability issues—like pit the connector power contacts, when hot-plugging a load such as a PCI card. The TPS2112A/3A slews the output voltage at a much faster rate when OUT switches between IN1 and IN2. The fast rate minimizes the output voltage droop and reduces the output voltage hold-up capacitance requirement.

#### SBVS045C - MARCH 2004 - REVISED MAY 2012



NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Cł	nanges from Revision B (March 2010) to Revision C	Page
•	Changed description of power supplies in <i>Description</i> section	
•	Changed Current Limit Adjustment Range parameter TPS2113A and TPS2115A specifications in Available Options table	
٠	Added I <sub>OUT</sub> column to Device Information table, changed table name	2
٠	Changed Continuous output current parameter in Absolute Maximum Ratings table	2
•	Changed Current limit adjustment range parameter in Recommended Operating Conditions table	3
•	Added footnote 1 to Recommended Operating Conditions table	3
•	Changed second paragraph in Application Information section	16

#### Changes from Revision A (February, 2006) to Revision B

#### Page

•	Updated document to current format	1
•	Deleted package information from Available Options table	2
•	Revised Ordering Information table	2
•	Deleted storage temperature, operating virtual junction temperature range, and lead temperature specifications from, added electrostatic discharge and junction temperature specifications to Absolute Maximum Ratings table; deleted ESD Protection table	2
•	Added DRB package information and footnote to <i>Dissipation Ratings</i> table	2



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11-Apr-2013

### PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
TPS2112APW	ACTIVE	TSSOP	PW	8	150	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	2112A	Samples
TPS2112APWG4	ACTIVE	TSSOP	PW	8	150	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	2112A	Samples
TPS2112APWR	ACTIVE	TSSOP	PW	8	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	2112A	Samples
TPS2112APWRG4	ACTIVE	TSSOP	PW	8	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	2112A	Samples
TPS2113ADRBR	ACTIVE	SON	DRB	8	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	PTOI	Samples
TPS2113ADRBT	ACTIVE	SON	DRB	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	PTOI	Samples
TPS2113APW	ACTIVE	TSSOP	PW	8	150	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	2113A	Samples
TPS2113APWG4	ACTIVE	TSSOP	PW	8	150	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	2113A	Samples
TPS2113APWR	ACTIVE	TSSOP	PW	8	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	2113A	Samples
TPS2113APWRG4	ACTIVE	TSSOP	PW	8	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	2113A	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)



11-Apr-2013

<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.

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# **PACKAGE MATERIALS INFORMATION**

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### **TAPE AND REEL INFORMATION**





### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal Device	1	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS2112APWR	TSSOP	PW	8	2000	330.0	12.4	7.0	3.6	1.6	8.0	12.0	Q1
TPS2113ADRBR	SON	DRB	8	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
TPS2113ADRBT	SON	DRB	8	250	180.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
TPS2113APWR	TSSOP	PW	8	2000	330.0	12.4	7.0	3.6	1.6	8.0	12.0	Q1

TEXAS INSTRUMENTS

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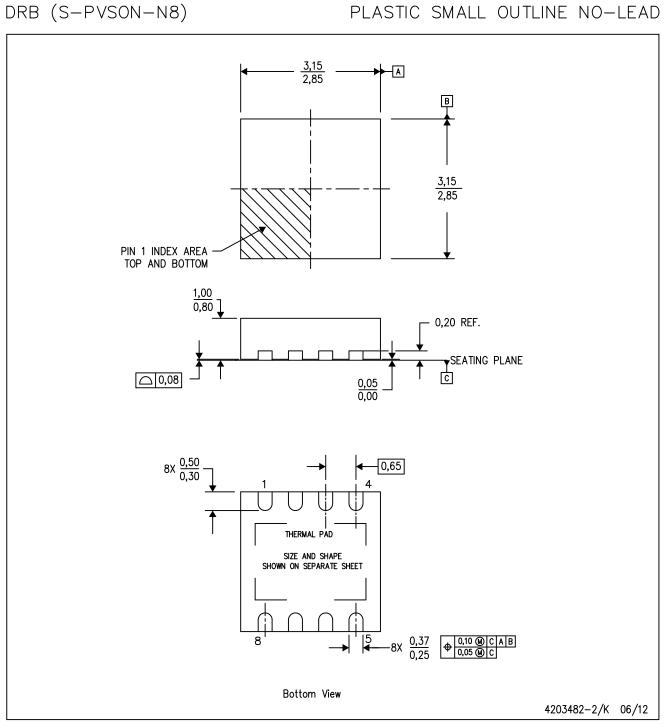
# PACKAGE MATERIALS INFORMATION

11-Jun-2013



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS2112APWR	TSSOP	PW	8	2000	367.0	367.0	35.0
TPS2113ADRBR	SON	DRB	8	3000	367.0	367.0	35.0
TPS2113ADRBT	SON	DRB	8	250	210.0	185.0	35.0
TPS2113APWR	TSSOP	PW	8	2000	367.0	367.0	35.0



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

- B. This drawing is subject to change without notice.
- C. Small Outline No-Lead (SON) package configuration.
- D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
- E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.



## THERMAL PAD MECHANICAL DATA

### DRB (S-PVSON-N8)

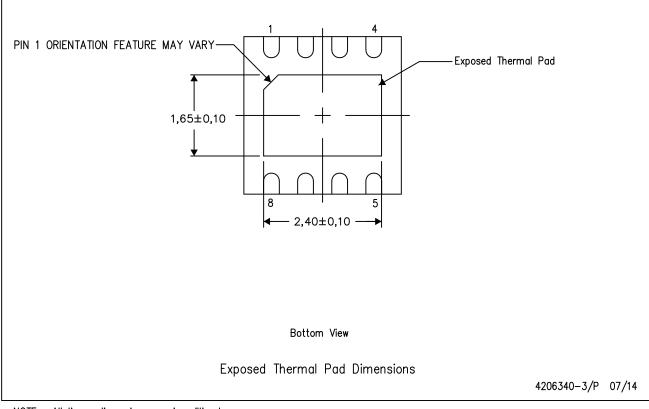
# PLASTIC SMALL OUTLINE NO-LEAD

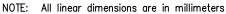
#### THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

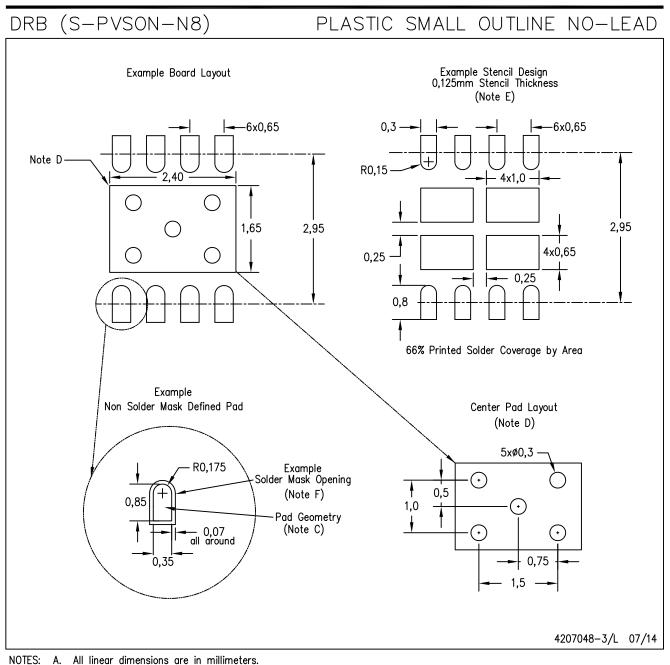
For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.









- A. All linear dimensions are in millimeters.
  B. This drawing is subject to change without notice.
  - C. Publication IPC-7351 is recommended for alternate designs.
  - D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, QFN Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at
- www.ti.com <http://www.ti.com>. E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should
- contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations. F. Customers should contact their board fabrication site for solder mask tolerances.



# **PW0008A**



# **PACKAGE OUTLINE**

### TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M. 2. This drawing is subject to change without notice. 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-153, variation AA.



# PW0008A

# **EXAMPLE BOARD LAYOUT**

### TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



# PW0008A

# **EXAMPLE STENCIL DESIGN**

### TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



NOTES: (continued)

9. Board assembly site may have different recommendations for stencil design.



<sup>8.</sup> Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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