**Low Noise Silicon Bipolar RF Transistor**

- Low noise amplifier designed for low voltage applications, ideal for 1.2 V or 1.8 V supply voltage
- Common e.g. in cordless phones, satellite receivers and oscillators up to 22 GHz
- High gain and low noise at high frequencies due to high transit frequency f_T = 45 GHz
- Easy to use Pb-free (RoHS compliant) and halogen free industry standard package with visible leads
- Qualification report according to AEC-Q101 available
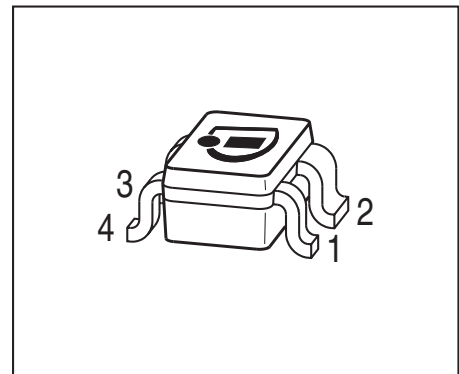
ESD (**E**lectro**s**tatic **d**ischarge) sensitive device, observe handling precaution!

Type	Marking	Pin Configuration						Package
BFP520	APs	1=B	2=E	3=C	4=E	-	-	SOT343

Maximum Ratings at T_A = 25 °C, unless otherwise specified

Parameter	Symbol	Value	Unit
Collector-emitter voltage	V_{CEO}		V
T_A = 25 °C		2.5	
T_A = -55 °C		2.4	
Collector-emitter voltage	V_{CES}	10	
Collector-base voltage	V_{CBO}	10	
Emitter-base voltage	V_{EBO}	1	
Collector current	I_C	50	mA
Base current	I_B	5	
Total power dissipation[1) $T_S \leq$ 105 °C	P_{tot}	125	mW
Junction temperature	T_J	150	°C
Storage temperature	T_{Stg}	-55 ... 150	

[1] T_S is measured on the emitter lead at the soldering point to pcb

Thermal Resistance

Parameter	Symbol	Value	Unit
Junction - soldering point[1]	R_{thJS}	450	K/W

Electrical Characteristics at T_A = 25 °C, unless otherwise specified

| Parameter	Symbol	Values			Unit
		min.	typ.	max.	
DC Characteristics					
Collector-emitter breakdown voltage I_C = 1 mA, I_B = 0	$V_{(BR)CEO}$	2.5	3	3.5	V
Collector-emitter cutoff current V_{CE} = 2 V, V_{BE} = 0 V_{CE} = 10 V, V_{BE} = 0	I_{CES}	- -	1 -	30 1000	nA
Collector-base cutoff current V_{CB} = 2 V, I_E = 0	I_{CBO}	-	-	30	
Emitter-base cutoff current V_{EB} = 0.5 V, I_C = 0	I_{EBO}	-	100	3000	
DC current gain I_C = 20 mA, V_{CE} = 2 V, pulse measured	h_{FE}	70	110	170	-

[1]For the definition of R_{thJS} please refer to Application Note AN077 (Thermal Resistance Calculation)

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**Electrical Characteristics** at T_A = 25 °C, unless otherwise specified

Parameter	Symbol	Values			Unit
		min.	typ.	max.	
AC Characteristics (verified by random sampling)					
Transition frequency I_C = 30 mA, V_{CE} = 2 V, f = 2 GHz	f_T	32	45	-	GHz
Collector-base capacitance V_{CB} = 2 V, f = 1 MHz, V_{BE} = 0 , emitter grounded	C_{cb}	-	0.07	0.13	pF
Collector emitter capacitance V_{CE} = 2 V, f = 1 MHz, V_{BE} = 0 , base grounded	C_{ce}	-	0.3	-	
Emitter-base capacitance V_{EB} = 0.5 V, f = 1 MHz, V_{CB} = 0 , collector grounded	C_{eb}	-	0.33	-	
Minimum noise figure I_C = 2 mA, V_{CE} = 2 V, Z_S = Z_{Sopt} , f = 1.8 GHz	NF_{min}	-	0.95	-	dB
Power gain, maximum stable[1] I_C = 20 mA, V_{CE} = 2 V, Z_S = Z_{Sopt}, Z_L = Z_{Lopt} , f = 1.8 GHz	G_{ms}	-	24	-	dB
Insertion power gain V_{CE} = 2 V, I_C = 20 mA, f = 1.8 GHz, Z_S = Z_L = 50 Ω	$|S_{21}|^2$	-	21.5	-	
Third order intercept point at output V_{CE} = 2 V, I_C = 20 mA, f = 1.8 GHz, Z_S = $Z_{Sopt,}$ Z_L = Z_{Lopt} V_{CE} = 2 V, I_C = 7 mA, f = 1.8 GHz, Z_S = $Z_{Sopt,}$ Z_L = Z_{Lopt}	IP_3	-	25 17	-	dBm
1dB compression point at output I_C = 20 mA, V_{CE} = 2 V, Z_S = $Z_{Sopt,}$ Z_L = Z_{Lopt}, f = 1.8 GHz I_C = 7 mA, V_{CE} = 2 V, Z_S = $Z_{Sopt,}$ Z_L = Z_{Lopt}, f = 1.8 GHz	P_{-1dB}	-	12 5	-	

[1]G_{ms} = $|S_{21} / S_{12}|$

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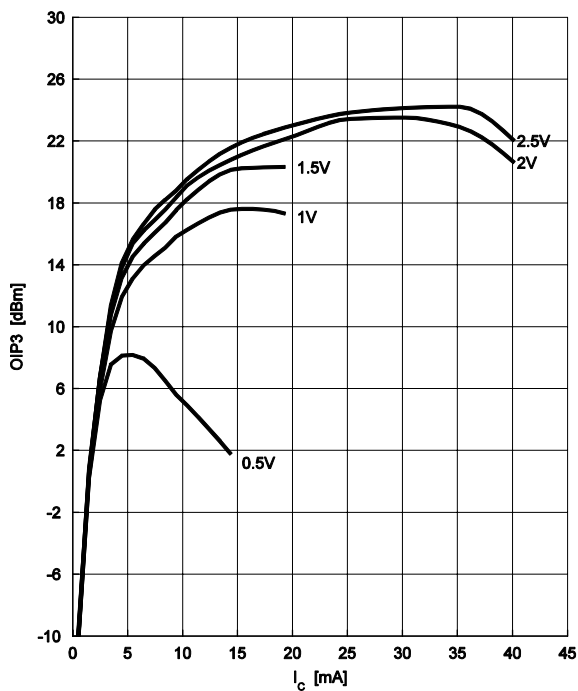
Total power dissipation $P_{tot} = f(T_S)$
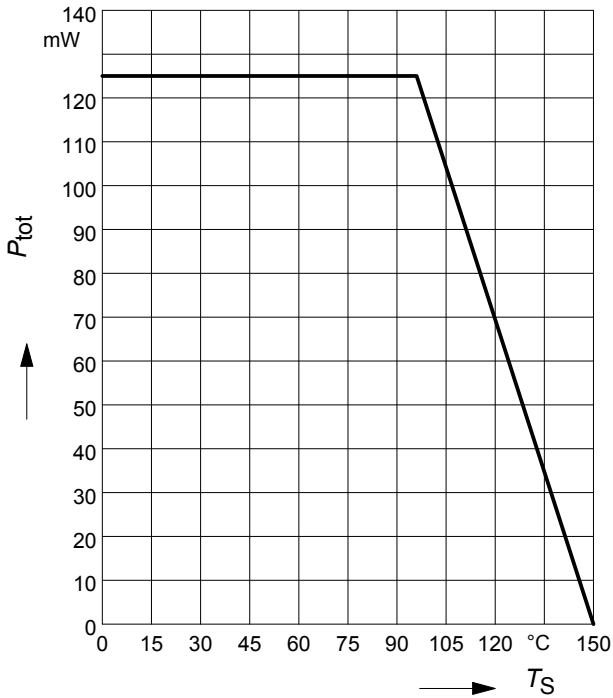
Collector-base capacitance $C_{cb} = f(V_{CB})$

f = 1MHz
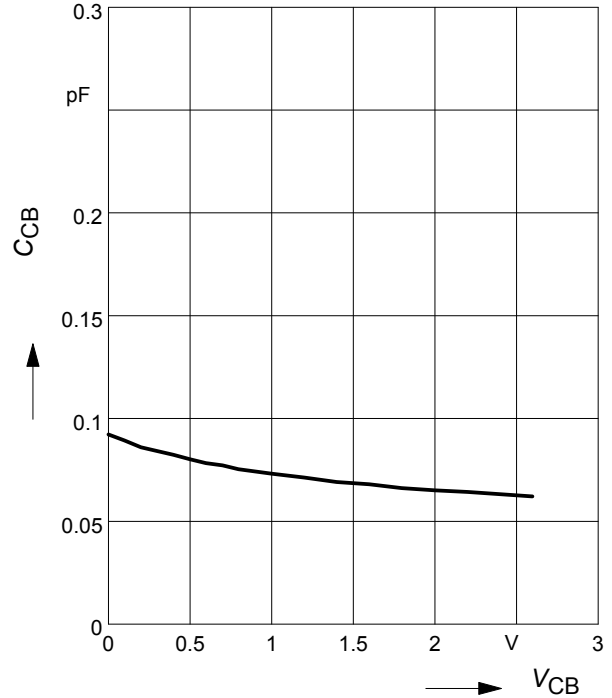
Third order Intercept Point $IP_3 = f(I_C)$

(Output, $Z_S = Z_L = 50\ \Omega$)
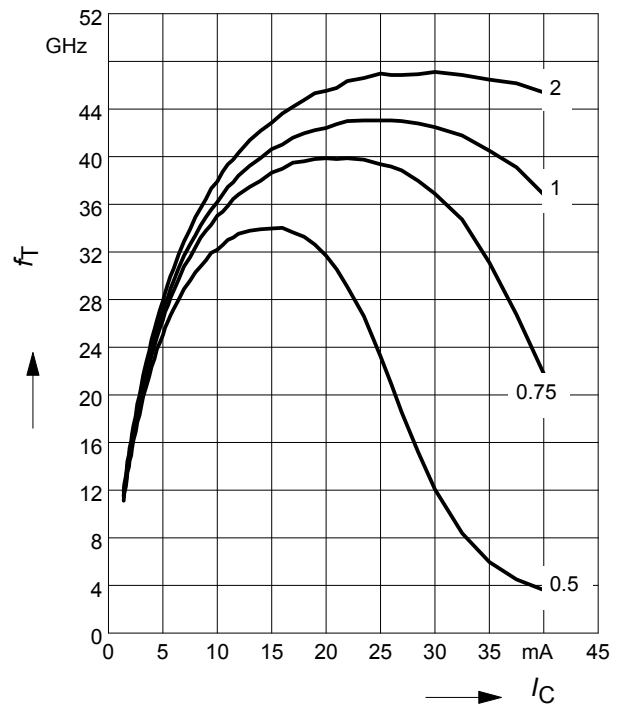
V_{CE} = parameter, f = 900 MHz
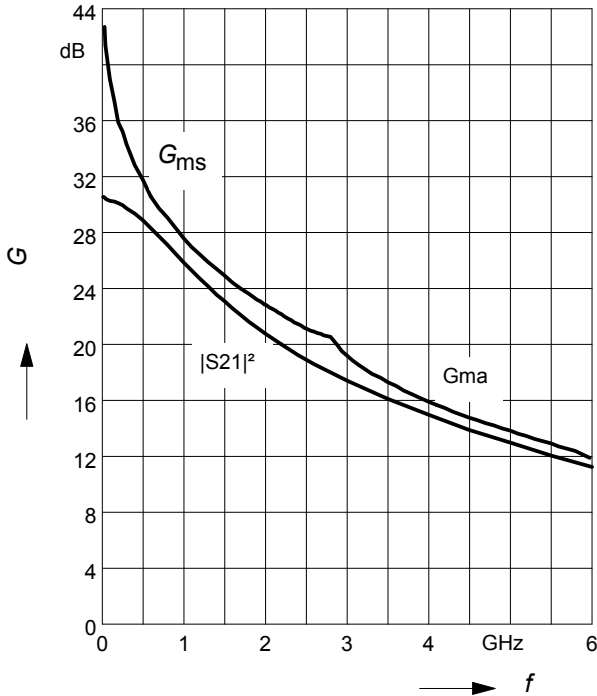

Transition frequency $f_T = f(I_C)$

f = 2 GHz

V_{CE} = parameter in V

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Power gain G_{ma}, G_{ms}, $|S_{21}|^2 = f(f)$

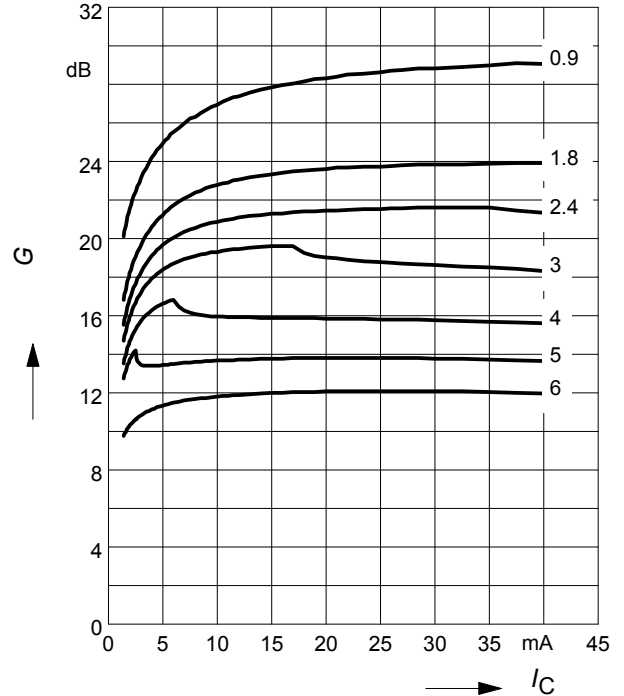$V_{CE} = 2$ V, $I_C = 20$ mA

Power gain G_{ma}, $G_{ms} = f(I_C)$

$V_{CE} = 2$V

f = parameter in GHz
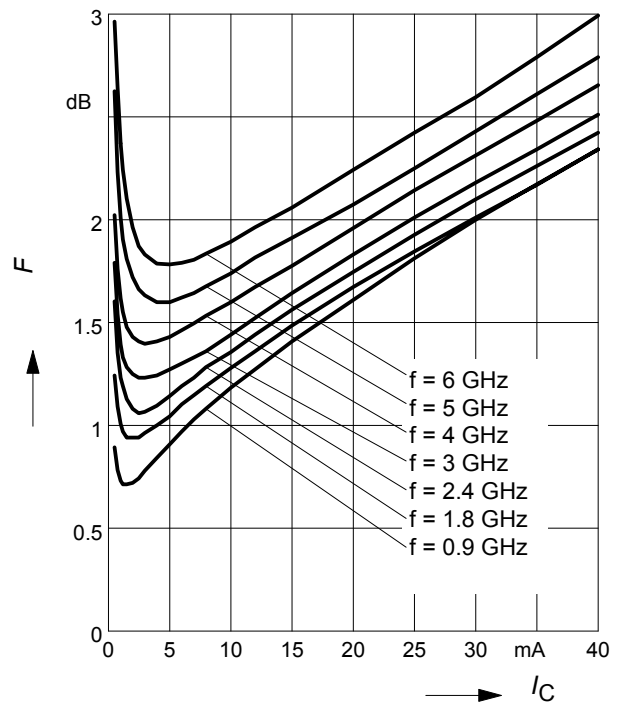

Power gain G_{ma}, $G_{ms} = f(V_{CE})$

$I_C = 20$ mA

f = parameter in GHz
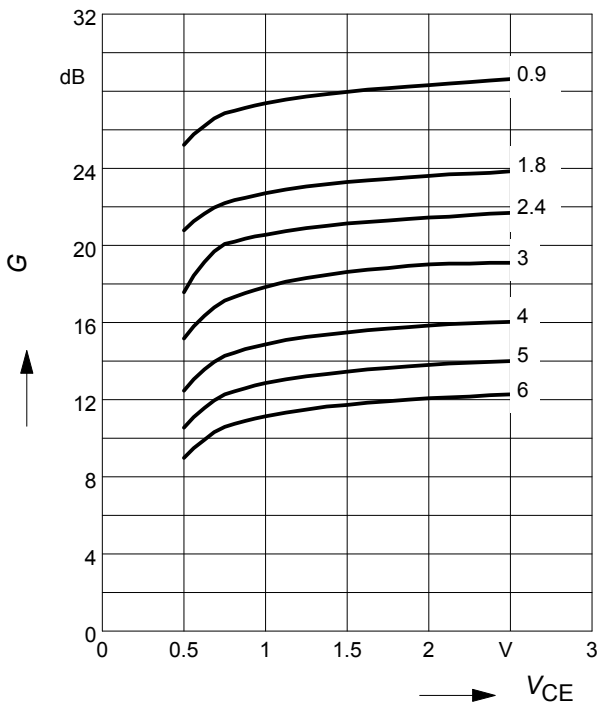
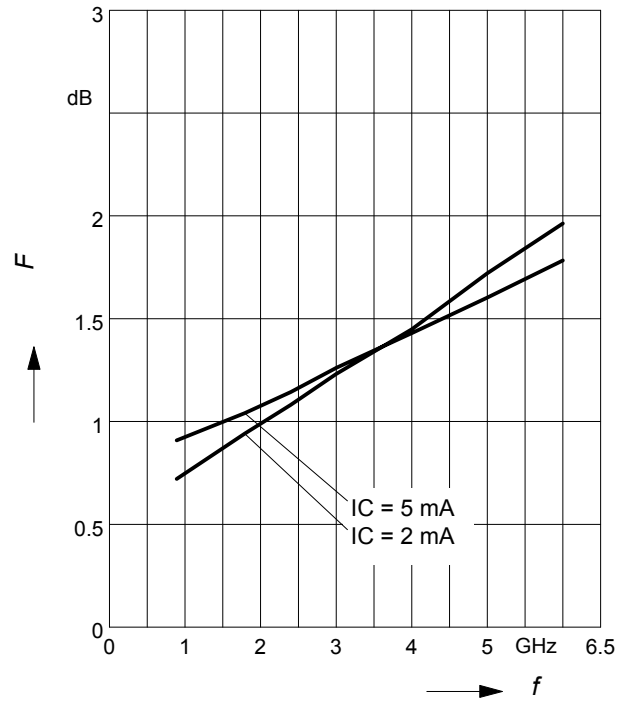**Minimum noise figure** $NF_{min} = f(I_C)$

$V_{CE} = 2$ V, $Z_S = Z_{Sopt}$

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Noise figure $F = f(I_C)$

V_{CE} = 2 V, f = 1.8 GHz
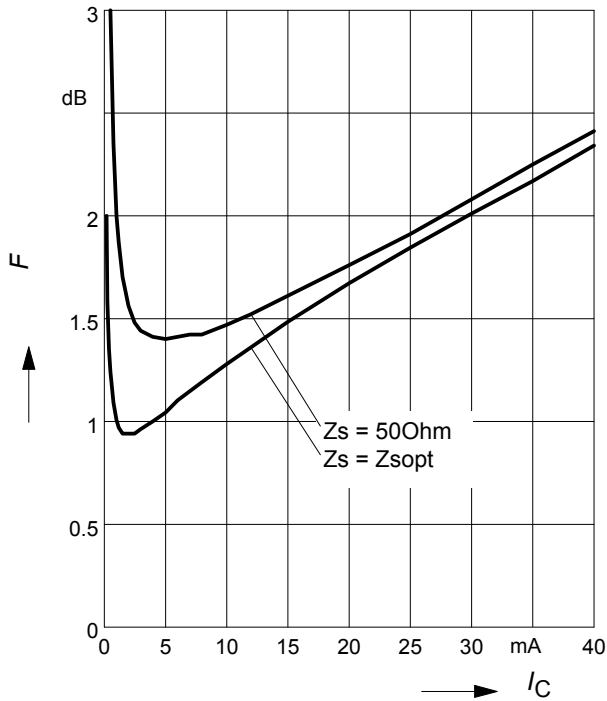
Minimum noise figure $NF_{min} = f(f)$

V_{CE} = 2 V, $Z_S = Z_{Sopt}$

Zs = 50Ohm
Zs = Zsopt

IC = 5 mA
IC = 2 mA

Source impedance for min.

noise figure vs. frequency

V_{CE} = 2 V, I_C = 2 mA / 5 mA
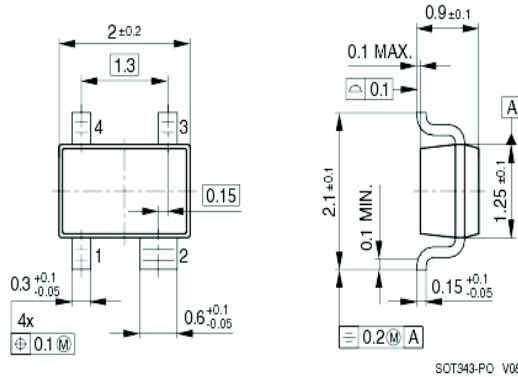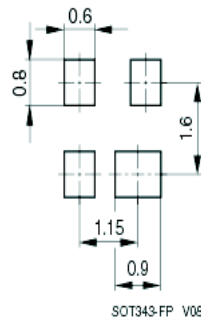
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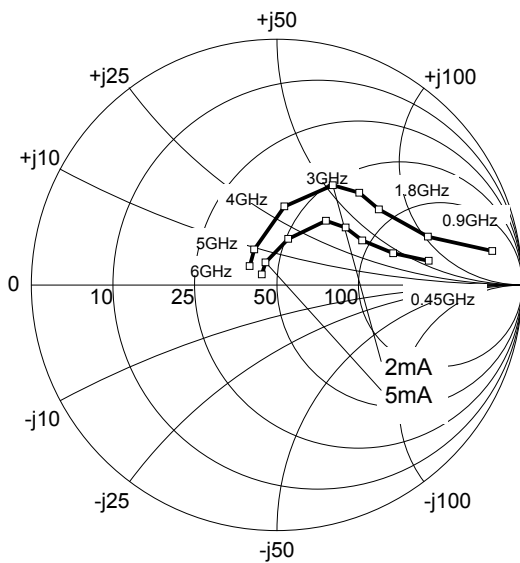
SPICE GP Model

For the SPICE Gummel Poon (GP) model as well as for the S-parameters
(including noise parameters) please refer to our internet website
www.infineon.com/rf.models.
Please consult our website and download the latest versions before actually
starting your design.You find the BFP520 SPICE GP model in the internet
in MWO- and ADS-format, which you can import into these circuit simulation tools
very quickly and conveniently. The model already contains the package parasitics
and is ready to use for DC and high frequency simulations. The terminals of the
model circuit correspond to the pin configuration of the device. The model
parameters have been extracted and verified up to 10 GHz using typical devices.
The BFP520 SPICE GP model reflects the typical DC- and RF-performance
within the limitations which are given by the SPICE GP model itself. Besides the DC
characteristics all S-parameters in magnitude and phase, as well as noise figure
(including optimum source impedance, equivalent noise resistance and flicker noise)
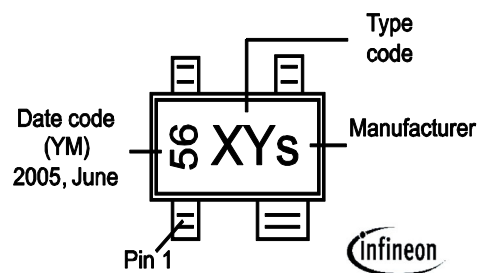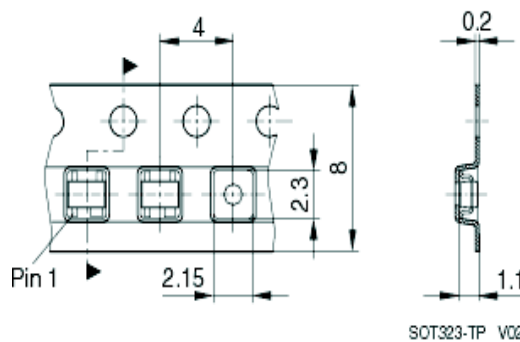and intermodulation have been extracted.

Package Outline


```
2±0.2
1.3
4        3
0.15
1        2
0.3 +0.1/-0.05
0.6 +0.1/-0.05
4x
⌖ 0.1 Ⓜ
```

```
0.9±0.1
0.1 MAX.
⌒ 0.1
A
2.1±0.1
0.1 MIN.
1.25 ±0.1
0.15 +0.1/-0.05
▤ 0.2 Ⓜ A
```

SOT343-PO V08

Foot Print


```
0.6
0.8
1.6
1.15
0.9
```

SOT343-FP V08

Marking Layout (Example)

Type code

Date code
(YM)
2005, June

56 XYs

Manufacturer

Pin 1

Standard Packing

Reel Ø180 mm = 3.000 Pieces/Reel
Reel Ø330 mm = 10.000 Pieces/Reel


```
4
8
2.3
Pin 1    2.15
```

```
0.2
1.1
```

SOT323-TP V02

Package SOT 343.vsd

Edition 2009-11-16

Information

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